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Adaptive λ -tracking for a class of infinite-dimensional systems¹

Achim Ilchmann^{a,2}, Hartmut Logemann^{b,*}

^a Department of Mathematics, University of Exeter, North Park Road, Exeter EX4 4QE, UK

^b Department of Mathematical Sciences, University of Bath, Claverton Down, Bath BA2 7AY, UK

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Abstract

For a class of high-gain stabilizable multivariable linear infinite-dimensional systems we present an adaptive control law which achieves approximate asymptotic tracking in the sense that the tracking error tends asymptotically to a ball centred at 0 and of arbitrary prescribed radius $\lambda > 0$. This control strategy, called λ -tracking, combines proportional error feedback with a simple nonlinear adaptation of the feedback gain. It does not involve any parameter estimation algorithms, nor is it based on the internal model principle. The class of reference signals is $W^{1,\infty}$, the Sobolev space of absolutely continuous functions which are bounded and have essentially bounded derivative. The control strategy is robust with respect to output measurement noise in $W^{1,\infty}$ and bounded input disturbances. We apply our results to retarded systems and integrodifferential systems. © 1998 Elsevier Science B.V. All rights reserved.

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1. Introduction

We present an adaptive regulator for the class of infinite-dimensional m -input– m -output systems described by

$$\dot{y}(t) = \mathcal{H}(y)(t) + Gu(t) + w(t), \quad y(0) = y_0 \in \mathbb{R}^m, \quad (1.1)$$

where as usual $u(\cdot)$ and $y(\cdot)$ denote the plant input and output, respectively, G is a real $m \times m$ -matrix whose eigenvalues have positive real parts, and \mathcal{H} is a causal linear operator, which is input–output stable in a certain sense; see Section 2 for details. In applications \mathcal{H} will be the input–output operator of a state-space system or a system described by a functional or partial differential equation. The function $w(\cdot)$ then models the effect of non-zero initial conditions. Our class covers retarded and integrodifferential systems which satisfy a generalized minimum phase condition and have a generalized high-frequency gain whose eigenvalues have positive real parts, and are hence stabilizable by static high-gain output feedback (for a detailed discussion see Section 4).

* Corresponding author. E-mail: hl@maths.bath.ac.uk.

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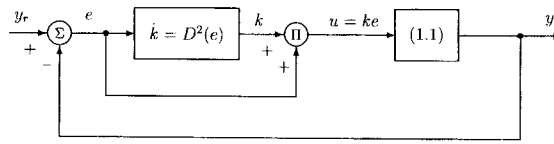


Fig. 1. Closed-loop system.

To the above class of systems we apply a simple adaptive controller of the form

$$\begin{aligned} e(t) &= y_r(t) - y(t), \\ u(t) &= k(t)e(t), \\ \dot{k}(t) &= \begin{cases} (\|e(t)\| - \lambda)^2 & \text{if } \|e(t)\| \geq \lambda, \\ 0 & \text{if } \|e(t)\| < \lambda \end{cases} \quad k(0) = k_0. \end{aligned} \quad (1.2)$$

This control strategy, called λ -tracking, is similar to a control law introduced by Ilchmann and Ryan [7] in a finite-dimensional context. In Eq. (1.2), $y_r(\cdot)$ is a reference signal which is assumed to belong to $W^{1,\infty}(\mathbb{R}_+, \mathbb{R}^m)$, i.e. y_r is absolutely continuous on every compact subinterval of \mathbb{R}_+ and y_r and \dot{y}_r are essentially bounded. The constant $\lambda > 0$ is an upper bound for the asymptotic tracking error and is chosen by the designer. Setting $D(e) = \|e\| - \lambda$ if $\|e\| \geq \lambda$ and $D(e) = 0$ otherwise, the gain adaptation in Eq. (1.2) can be written as $\dot{k} = D^2(e)$. The closed-loop system given by Eqs. (1.1) and (1.2) is shown in Fig. 1.

Our main result shows that for all reference signals $y_r \in W^{1,\infty}(\mathbb{R}_+, \mathbb{R}^m)$, all initial conditions $(y_0, k_0) \in \mathbb{R}^m \times \mathbb{R}$ and all essentially bounded functions $w(\cdot)$, the controller (1.2) achieves convergence of the feedback gain $k(t)$, and the output $y(t)$ will approach the ball $B_\lambda(y_r(t))$ of radius λ centred at $y_r(t)$ as $t \rightarrow \infty$, i.e.

$$\limsup_{t \rightarrow \infty} \|e(t)\| \leq \lambda. \quad (1.3)$$

Moreover, we show that the closed-loop system is robust with respect to measurement noise in $W^{1,\infty}(\mathbb{R}_+, \mathbb{R}^m)$ and bounded input disturbances. Note that the control law (1.2) is of striking simplicity – it combines proportional error feedback with a simple nonlinear adaptation of the feedback gain. It does not invoke any parameter estimation, nor is it based on the internal model principle. We mention that for finite-dimensional problems, the concept of λ -tracking has been successfully applied to the control of industrial plants and processes as for example: continuous stirred tank reactors for methanol synthesis [1], binary distillation columns [2] and biogas tower reactors [6].

The technique of representing a large class of high-gain stabilizable infinite-dimensional linear systems by an abstract Volterra integrodifferential equation of the form (1.1) was introduced by Logemann and Owens [9]. It has been exploited for adaptive asymptotic tracking by Logemann and Ilchmann [8]. However, in [8] we needed to invoke the internal model principle and the class of reference signals was essentially restricted to finite sums of sinusoids. In this paper we overcome this drawback by weakening the control objective slightly: instead of exact asymptotic tracking the design goal is approximate asymptotic tracking in the sense of Eq. (1.3).

The paper is organized as follows. Section 2 contains some preliminaries on abstract Volterra integrodifferential systems which are needed to establish existence and uniqueness of solutions for the closed-loop system given by Eqs. (1.1) and (1.2). Moreover, we give a frequency-domain interpretation for our class of systems in the case of shift-invariant \mathcal{H} . Section 3 contains the main result on adaptive λ -tracking, whilst Section 4 is devoted to the application of this result to retarded and integrodifferential systems.

Nomenclature. As usual, set $\mathbb{C}_\alpha := \{s \in \mathbb{C} \mid \operatorname{Re} s > \alpha\}$, where $\alpha \in \mathbb{R}$. The Euclidean norm on \mathbb{R}^n and \mathbb{C}^n and the matrix norm induced by the Euclidean norm will be denoted by $\|\cdot\|$. We will make use of the following function spaces, where $I \subset \mathbb{R}$ denotes an interval and $\mathbb{K} = \mathbb{R}, \mathbb{C}$.

- $C(I, \mathbb{K}^n)$:= vector space of all continuous functions on I with values in \mathbb{K}^n .
- $L^p(I, \mathbb{K}^n)$:= vector space of all p -integrable functions on I with values in \mathbb{K}^n , $p \geq 1$.
- $L^\infty(I, \mathbb{K}^n)$:= vector space of all essentially bounded functions on I with values in \mathbb{K}^n .
- $L^p_{\text{loc}}(I, \mathbb{K}^n)$:= vector space of all locally p -integrable functions on I with values in \mathbb{K}^n , $p \geq 1$.
- $W^{1,\infty}(\mathbb{R}_+, \mathbb{K}^n)$:= vector space of all \mathbb{K}^n -valued functions f defined on \mathbb{R}_+ such that f is absolutely continuous on every compact subinterval of \mathbb{R}_+ and f and \dot{f} are essentially bounded.
- $BV(I, \mathbb{K}^{n \times n})$:= vector space of $\mathbb{K}^{n \times n}$ -valued functions of bounded variation defined on I .
- $M(\mathbb{R}_+, \mathbb{K}^{n \times n})$:= vector space of bounded Borel measures on \mathbb{R}_+ with values in $\mathbb{K}^{n \times n}$.
- $H^\infty(\mathbb{C}^{n \times n})$:= algebra of bounded holomorphic functions defined on \mathbb{C}_0 with values in $\mathbb{C}^{n \times n}$.

Let f be a function defined on $[0, a]$, where $0 < a \leq \infty$. Then for all $\tau \in [0, a]$

$$(\mathcal{P}_\tau f)(t) := \begin{cases} f(t), & 0 \leq t \leq \tau, \\ 0, & t > \tau. \end{cases}$$

If X is a normed space, then $B(X)$ denotes the set of linear bounded operators from X into X . If $T \in B(X)$, then $\sigma(T)$ denotes the spectrum of T .

\mathcal{L} denotes the Laplace transform. The superscript $\hat{\cdot}$ is used to denote Laplace transformed or Laplace–Stieltjes transformed functions.

2. Preliminaries and system description

The plant to be controlled is given by Eq. (1.1), where we assume that

- (A1) $G \in \mathbb{R}^{m \times m}$ with $\sigma(G) \subset \mathbb{C}_0$,
- (A2) \mathcal{H} is causal and $\mathcal{H} \in B(L^2(\mathbb{R}_+, \mathbb{R}^m)) \cap B(L^\infty(\mathbb{R}_+, \mathbb{R}^m))$.

Recall that \mathcal{H} is called shift-invariant if $\mathcal{S}_t \mathcal{H} = \mathcal{H} \mathcal{S}_t$ for all $t \geq 0$, where \mathcal{S}_t denotes the operator of right-shift by t . Since shift-invariance implies causality (see [11]), assumption (A2) is implied by

- (A2') \mathcal{H} is shift-invariant and $\mathcal{H} \in B(L^2(\mathbb{R}_+, \mathbb{R}^m)) \cap B(L^\infty(\mathbb{R}_+, \mathbb{R}^m))$.

Assumption (A2') (and hence assumption (A2)) is usually satisfied for the input–output operators of systems given by linear autonomous exponentially stable differential equations (ODEs, PDEs as well as FDEs). In particular, it is satisfied for the classes of retarded and integrodifferential systems considered in Section 4.

It follows from [12] that (A2') holds if and only if \mathcal{H} is a convolution operator of the form $\mathcal{H}(y) = H * y$, where $H \in M(\mathbb{R}_+, \mathbb{R}^{m \times m})$. In this case, if $y_0 = 0$ and $w \equiv 0$, Laplace transformation of Eq. (1.1) gives

$$s \hat{y}(s) = \hat{H}(s) \hat{y}(s) + G \hat{u}(s).$$

Setting

$$G(s) = (sI - \hat{H}(s))^{-1} G, \tag{2.1}$$

it follows that $\hat{y}(s) = G(s) \hat{u}(s)$. Using assumption (A1) and the fact that \hat{H} is in $H^\infty(\mathbb{C}^{m \times m})$, it is not difficult to show that for all sufficiently large $\gamma > 0$

$$G(I + \gamma G)^{-1} \in H^\infty(\mathbb{C}^{m \times m}),$$

i.e. static high-gain feedback leads to a L^2 -stable closed-loop system.³ This observation is the motivation for applying the high-gain adaptive control law (1.2) to the system (1.1).

³ In fact, combining this result with a Paley–Wiener type theorem for integrodifferential equations (see Theorem 3.5 on p. 83 in [3]), it can be shown that for sufficiently large $\gamma > 0$, the inverse Laplace transform of $G(I + \gamma G)^{-1}$ is in $L^1(\mathbb{R}_+, \mathbb{R}^{m \times m})$. Hence it follows that the closed-loop system is L^p -stable for all $p \in [1, \infty]$.

Conversely, if a system is externally described by a transfer function $G(s)$, and there exists $\hat{H} \in H^\infty(\mathbb{C}^{m \times m})$ and $G \in \mathbb{R}^{m \times m}$ with $\sigma(G) \subset \mathbb{C}_0$ such that (2.1) holds, then the system can be represented by Eq. (1.1), where $\mathcal{H}(y) = \mathfrak{L}^{-1}(\hat{H}\hat{y})$. Whilst in this case the operator \mathcal{H} is shift-invariant and in $B(L^2(\mathbb{R}_+, \mathbb{R}^m))$, it is not true in general that $\mathcal{H} \in B(L^\infty(\mathbb{R}_+, \mathbb{R}^m))$. However, under the extra assumption that the inverse Laplace transform H of \hat{H} is in $M(\mathbb{R}_+, \mathbb{R}^{m \times m})$ (which is usually satisfied in applications), it follows that $\mathcal{H} \in B(L^\infty(\mathbb{R}_+, \mathbb{R}^m))$.

Finally, we mention that a transfer function G which is meromorphic in \mathbb{C}_α for some $\alpha < 0$, admits a factorization of the form (2.1) with G nonsingular and $\hat{H} \in H^\infty(\mathbb{C}^{m \times m})$ if and only if $sG(s) - G = O(s)$ as $|s| \rightarrow \infty$ in \mathbb{C}_0 (generalized “relative-degree-one” condition) and $G(s)$ has no zeros in \mathbb{C}_0^{cl} (minimum-phase condition), see [10].

To prove existence and uniqueness of the solution of the closed-loop system given by Eqs. (1.1) and (1.2), we first consider the following more general initial-value problem

$$\dot{x}(t) = (\mathcal{A}x)(t) + f(t, x(t)) + g(t), \quad t \geq 0, \quad (2.2a)$$

$$x(0) = x_0 \in \mathbb{R}^n. \quad (2.2b)$$

Here we assume:

- (i) $\mathcal{A} : L^2_{loc}(\mathbb{R}_+, \mathbb{R}^n) \rightarrow L^2_{loc}(\mathbb{R}_+, \mathbb{R}^n)$, $\mathcal{A}(0) = 0$ and there exists $\kappa > 0$ such that

$$\|\mathcal{P}_t(\mathcal{A}x - \mathcal{A}x')\|_{L^2} \leq \kappa \|\mathcal{P}_t(x - x')\|_{L^2} \quad \text{for all } x, x' \in L^2_{loc}(\mathbb{R}_+, \mathbb{R}^n), t \geq 0,$$

i.e. \mathcal{A} is unbiased, causal and of finite incremental gain;

- (ii) $f : \mathbb{R}_+ \times \mathbb{R}^n \rightarrow \mathbb{R}^n$ is a function with $f(t, x)$ being continuous in t and locally Lipschitz continuous in x , uniformly in t on bounded intervals;

- (iii) g is in $L^1_{loc}(\mathbb{R}_+, \mathbb{R}^n)$.

In order to define what we mean by a solution of the initial value problem (2.2) on $[0, T)$, where $0 < T \leq \infty$, we have to give a meaning to $\mathcal{A}x$ if $x \in C([0, T), \mathbb{R}^n)$ (remember that \mathcal{A} operates on functions whose domain of definition is \mathbb{R}_+). We set $(\mathcal{A}x)(t) = (\mathcal{A}\mathcal{P}_\tau x)(t)$ for $0 \leq t \leq \tau < T$. Since \mathcal{A} is causal, this definition does not depend on the choice of τ . By a *solution* of Eqs. (2.2a) and (2.2b) on $[0, T)$ we mean a function x defined on $[0, T)$ which is absolutely continuous on every compact subinterval of $[0, T)$, satisfies the differential equation (2.2a) for almost every $t \in [0, T)$ and matches the initial condition (2.2b).

Theorem 2.1. *The initial-value problem (2.2) has a unique solution defined on a maximal interval of existence $[0, \omega)$, where $0 < \omega \leq \infty$. If $\omega < \infty$, then there exists a sequence $t_i \in (0, \omega)$ with $\lim_{i \rightarrow \infty} t_i = \omega$ and such that $\lim_{i \rightarrow \infty} \|x(t_i)\| = \infty$.*

The above theorem has been proved in [9]. Similar results can be found in [3] (p. 359) and in [4]. Theorem 2.1 implies that the initial value problem (1.1) has a unique solution for all $u, w \in L^1_{loc}(\mathbb{R}_+, \mathbb{R}^m)$ and $y_0 \in \mathbb{R}^m$. Moreover, it shows that the closed-loop system given by Eqs. (1.1) and (1.2) has a unique solution.

3. Adaptive λ -tracking

The following theorem is the main result of this paper.

Theorem 3.1. *Let the assumptions (A1) and (A2) be satisfied and let $\lambda > 0$ be given. Then, for all initial conditions $y_0 \in \mathbb{R}^m$, $k_0 \in \mathbb{R}$, all $w \in L^\infty(\mathbb{R}_+, \mathbb{R}^m)$ and all reference signals $y_r \in W^{1, \infty}(\mathbb{R}_+, \mathbb{R}^m)$, the closed-loop system given by Eqs. (1.1) and (1.2) has the following properties:*

- (1) *The unique solution $(y(\cdot), k(\cdot))$ exists on $[0, \infty)$,*
- (2) *$\lim_{t \rightarrow \infty} k(t)$ exists and is finite,*
- (3) *$\limsup_{t \rightarrow \infty} \|y_r(t) - y(t)\| \leq \lambda$.*

Proof. To reduce the technical effort, we assume without loss of generality that $k_0 \geq 0$ in Eq. (1.2). By assumption (A1) there exists a positive definite real $m \times m$ -matrix $P = P^T$ satisfying the Lyapunov equation

$$G^T P + P G = 2I. \quad (3.1)$$

We define the P -induced norm $\|\cdot\|_P$ on \mathbb{R}^m by setting $\|x\|_P = \sqrt{\langle x, P x \rangle}$. Let $\varrho > 0$ denote the square root of the smallest eigenvalue of P and set $\varrho_P = 1/\sqrt{\|P\|}$. Then,

$$\varrho \|x\| \leq \|x\|_P, \quad \varrho_P \|x\|_P \leq \|x\| \quad \text{for all } x \in \mathbb{R}^m. \quad (3.2)$$

We introduce the following functions which will be used throughout the proof:

$$D : \mathbb{R}^m \rightarrow \mathbb{R}_+, \quad x \mapsto D(x) = \begin{cases} \|x\| - \lambda & \text{if } \|x\| \geq \lambda \\ 0 & \text{if } \|x\| < \lambda, \end{cases}$$

$$d : \mathbb{R}^m \rightarrow \mathbb{R}^m, \quad x \mapsto d(x) = \begin{cases} \frac{\|x\| - \lambda}{\|x\|} x & \text{if } \|x\| \geq \lambda \\ 0 & \text{if } \|x\| < \lambda, \end{cases}$$

$$D_P : \mathbb{R}^m \rightarrow \mathbb{R}, \quad x \mapsto D_P(x) = \begin{cases} \|x\|_P - \lambda \varrho & \text{if } \|x\|_P \geq \lambda \varrho \\ 0 & \text{if } \|x\|_P < \lambda \varrho, \end{cases}$$

$$d_P : \mathbb{R}^m \rightarrow \mathbb{R}^m, \quad x \mapsto d_P(x) = \begin{cases} \frac{\|x\|_P - \lambda \varrho}{\|x\|_P} x & \text{if } \|x\|_P \geq \lambda \varrho \\ 0 & \text{if } \|x\|_P < \lambda \varrho. \end{cases}$$

The function D is the same as introduced in Section 1. Clearly, for given $x \in \mathbb{R}^m$, $D(x)$ (respectively, $D_P(x)$) is the distance of x from the ball of radius λ (respectively, $\lambda \varrho$) centred at 0 in the norm $\|\cdot\|$ (respectively, $\|\cdot\|_P$).

We proceed in five steps.

Step 1 (Existence and uniqueness of a maximal solution): Setting

$$e = y_r - y \quad \text{and} \quad \tilde{w} = \dot{y}_r - w - \mathcal{H}(y_r)$$

the closed-loop system given by Eqs. (1.1) and (1.2) can be written in the following form:

$$\dot{e}(t) = -k(t)Ge(t) + \mathcal{H}(e)(t) + \tilde{w}(t), \quad e(0) = e_0 := y_r(0) - y_0, \quad (3.3a)$$

$$\dot{k}(t) = D^2(e(t)), \quad k(0) = k_0. \quad (3.3b)$$

By assumption $y_r \in W^{1,\infty}(\mathbb{R}_+, \mathbb{R}^m)$, and so $\dot{y}_r \in L^\infty(\mathbb{R}_+, \mathbb{R}^m)$ and, by assumption (A2), $\mathcal{H}(y_r) \in L^\infty(\mathbb{R}_+, \mathbb{R}^m)$. Moreover, $w \in L^\infty(\mathbb{R}_+, \mathbb{R}^m)$, and thus

$$\tilde{w} \in L^\infty(\mathbb{R}_+, \mathbb{R}^m). \quad (3.4)$$

The system (3.3) is of the form (2.2), and hence we can apply Theorem 2.1 to conclude that Eqs. (3.3a) and (3.3b) has a unique maximal solution $(e(\cdot), k(\cdot))$ defined on $[0, \omega)$, the maximal interval of existence. Using the linearity of \mathcal{H} and Eq. (3.3a), we see that e satisfies

$$\dot{e}(t) = -k(t)Ge(t) + \mathcal{H}(d_P(e))(t) + h(t), \quad (3.5)$$

where

$$h(t) := \mathcal{H}(e - d_P(e))(t) + \tilde{w}(t), \quad t \in [0, \omega).$$

It follows from the definition of d_P that $e - d_P(e) \in L^\infty(0, \omega; \mathbb{R}^m)$ and hence by assumption (A2) and Eq. (3.4) we conclude

$$h \in L^\infty(0, \omega; \mathbb{R}^m). \quad (3.6)$$

Step 2 (An estimate for a Lyapunov-type function): We claim that there exist constants $\mu_1, \mu_2 > 0$ such that differentiation of the function

$$V : \mathbb{R}^m \rightarrow \mathbb{R}_+, \quad x \mapsto V(x) = \frac{1}{2} D_P^2(x)$$

along the solution of Eq. (3.5) yields, for almost all $t \in [0, \omega)$,

$$\frac{d}{dt} V(e(t)) \leq -\mu_1 [k(t) - \mu_2] D_P(e(t)) \|e(t)\|_P + \mu_2 \|d_P(e(t))\| \| \mathcal{H}(d_P(e))(t) \|. \quad (3.7)$$

In the following the argument t will be omitted for brevity. By Eqs. (3.1), (3.2), (3.5) and (3.6) and since $k(t) \geq k_0 \geq 0$, we have almost everywhere on $[0, \omega)$,

$$\begin{aligned} \frac{d}{dt} V(e) &= \frac{D_P(e)}{\|e\|_P} \langle e, P\dot{e} \rangle \\ &= -\frac{D_P(e)}{\|e\|_P} k \|e\|^2 + \frac{D_P(e)}{\|e\|_P} \langle e, P\mathcal{H}(d_P(e)) + Ph \rangle \\ &\leq -\varrho^2 k D_P(e) \|e\|_P + \|P\| \|d_P(e)\| \| \mathcal{H}(d_P(e)) \| + \|P\| \|h\|_{L^\infty(0, \omega)} \|d_P(e)\|. \end{aligned} \quad (3.8)$$

Using Eq. (3.2) and the definitions of D_P and d_P we obtain

$$\|d_P(x)\| \leq D_P(x) \frac{1}{\varrho} \leq D_P(x) \frac{1}{\varrho^2 \lambda} \|x\|_P \quad \text{for all } x \in \mathbb{R}^m. \quad (3.9)$$

Combining Eqs. (3.8) and (3.9), we see that there exist positive constants μ_1 and μ_2 such that Eq. (3.7) holds on $[0, \omega)$.

Step 3 (Boundedness of k on $[0, \omega)$): We claim that $k \in L^\infty(0, \omega; \mathbb{R})$. Let $\kappa > 0$ be the L^2 -induced operator norm of \mathcal{H} . It follows from Hölder's inequality that for all $f \in L^2_{\text{loc}}(\mathbb{R}_+, \mathbb{R}^m)$ and all $t \in \mathbb{R}_+$,

$$\int_0^t \|f\| \| \mathcal{H}(f) \| \leq \kappa \int_0^t \|f\|^2. \quad (3.10)$$

Now either $k(t) \leq \mu_2(1 + \kappa/\mu_1\varrho^2)$ for all $t \in [0, \omega)$, whence the claim follows, or there exists a $t_0 \in [0, \omega)$ such that $k(t_0) > \mu_2(1 + \kappa/\mu_1\varrho^2)$. Then

$$\mu_3 := \mu_1(k(t_0) - \mu_2) > \kappa\mu_2/\varrho^2 > 0. \quad (3.11)$$

Integration of Eq. (3.7) from t_0 to $t \in [0, \omega)$, using that $k(\cdot)$ is nondecreasing and an application of Eq. (3.10) yields

$$\begin{aligned} V(e(t)) &\leq V(e(t_0)) - \mu_3 \int_{t_0}^t D_P(e) \|e\|_P + \mu_2 \int_{t_0}^t \|d_P(e)\| \| \mathcal{H}(d_P(e)) \| \\ &\leq V(e(t_0)) - \mu_3 \int_{t_0}^t D_P(e) \|e\|_P + \kappa\mu_2 \int_0^t \|d_P(e)\|^2. \end{aligned} \quad (3.12)$$

Clearly, for all $x \in \mathbb{R}^m$, we have

$$D_P(x) \|x\|_P = D_P(x) (\|x\|_P - \varrho\lambda) + \varrho\lambda D_P(x) \geq D_P^2(x).$$

Combining this with Eqs. (3.2) and (3.12) shows that

$$0 \leq V(e(t_0)) + \kappa\mu_2 \int_0^{t_0} \|d_P(e)\|^2 - \int_{t_0}^t (\mu_3 - \kappa\mu_2/\varrho^2) D_P^2(e).$$

Setting

$$\mu_4 := V(e(t_0)) + \kappa\mu_2 \int_0^{t_0} \|d_P(e)\|^2, \quad \mu_5 := 2(\mu_3 - \kappa\mu_2/\varrho^2),$$

we obtain

$$0 \leq \mu_4 - \mu_5 \int_{t_0}^t V(e) \quad \text{for all } t \in [t_0, \omega).$$

By Eq. (3.11), $\mu_5 > 0$, and thus we may conclude

$$\int_{t_0}^t V(e) \leq \mu_4/\mu_5 \quad \text{for all } t \in [t_0, \omega). \tag{3.13}$$

Now, using Eq. (3.2), it is easy to show that

$$D_P(x) \geq \varrho D(x) \quad \text{for all } x \in \mathbb{R}^m.$$

Therefore, for all $t \in [t_0, \omega)$,

$$k(t) = k(t_0) + \int_{t_0}^t D^2(e) \leq k(t_0) + \frac{1}{\varrho^2} \int_{t_0}^t D_P^2(e).$$

Appealing to Eq. (3.13), we obtain

$$k(t) \leq k(t_0) + \frac{2\mu_4}{\mu_5\varrho^2} \quad \text{for all } t \in [t_0, \omega),$$

which shows that $k(\cdot)$ is bounded on $[0, \omega)$.

Step 4 (Boundedness of e on $[0, \omega)$): From Eq. (3.13) we see that $D_P(e) \in L^2(0, \omega; \mathbb{R})$, and thus $d_P(e) \in L^2(0, \omega; \mathbb{R}^m)$. Moreover, by the definition of the function $d_P(\cdot)$ we have that $e - d_P(e) \in L^\infty(0, \omega; \mathbb{R}^m)$. Hence Step 3, assumption (A2) and Eq. (3.6) yield

$$\begin{aligned} f_1 &:= (1 - k)Gd_P(e) + \mathcal{H}(d_P(e)) \in L^2(0, \omega; \mathbb{R}^m), \\ f_2 &:= (1 - k)G[e - d_P(e)] + h \in L^\infty(0, \omega; \mathbb{R}^m). \end{aligned}$$

Trivially, Eq. (3.5) may be rewritten as

$$\dot{e}(t) = -Ge(t) + f_1(t) + f_2(t).$$

An application of the variations-of-constants formula to this equation yields

$$e(t) = e^{-Gt}e_0 + \int_0^t e^{-G(t-\tau)}[f_1(\tau) + f_2(\tau)]d\tau.$$

By assumption (A1), e^{-Gt} is exponentially stable, and therefore by a standard result on convolutions (see Theorem 2.2 on p. 39 in [3]) we obtain that

$$e \in L^\infty(0, \omega; \mathbb{R}^m). \tag{3.14}$$

Step 5 (Global existence and convergence): From Steps 3 and 4 we have boundedness of k and e on $[0, \omega)$. Therefore, combining the maximality of ω and Theorem 2.1, we obtain that $\omega = \infty$, and hence statement (1) follows. Consequently, using Steps 3 and 4, we conclude that $k(\cdot)$ and $e(\cdot)$ are bounded on $[0, \infty)$. Since $k(\cdot)$ is non-decreasing we obtain statement (2). To prove statement (3), note that by assumption (A2), Eqs. (3.5) and (3.6) and the boundedness of e on $[0, \infty)$,

$$\dot{e} \in L^\infty(\mathbb{R}_+, \mathbb{R}^m). \tag{3.15}$$

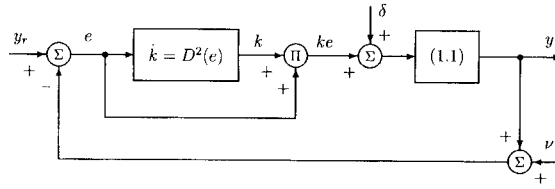


Fig. 2. Closed-loop system with input disturbances and measurement noise.

An easy calculation leads to

$$\left| \frac{d}{dt} D^2(e) \right| = 2 \frac{D(e)}{\|e\|} |\langle e, \dot{e} \rangle| \leq 2D(e) \|\dot{e}\|, \quad (3.16)$$

so that from Eqs. (3.14)–(3.16) we obtain

$$\frac{d}{dt} D^2(e) \in L^\infty(\mathbb{R}_+, \mathbb{R}). \quad (3.17)$$

Moreover, since $k \in L^\infty(\mathbb{R}_+, \mathbb{R})$, we have

$$D^2(e) \in L^1(\mathbb{R}_+, \mathbb{R}). \quad (3.18)$$

Finally, Eqs. (3.17) and (3.18) show that (see e.g. Lemma 2.1.7 in [5])

$$\lim_{t \rightarrow \infty} D^2(e(t)) = 0,$$

which in turn implies statement (3). \square

The following remark shows that the controller (1.2) is robust with respect to measurement noise and input disturbances.

Remark 3.2. (1) Suppose that the feedback system is subject to a bounded input disturbance δ and $W^{1,\infty}$ -measurement noise ν , see Fig. 2. This means we have to replace e and u in (1.2) by $e(t) = y_r(t) - y(t) - \nu(t)$ and $u(t) = k(t)e(t) + \delta(t)$, respectively. Then, by absorbing δ into the \tilde{w} term (see Step 1 in the proof of Theorem 3.1), it follows that statements (1) and (2) of Theorem 3.1 remain true and, moreover

$$\limsup_{t \rightarrow \infty} \|y_r(t) - y(t) - \nu(t)\| \leq \lambda, \quad (3.19)$$

provided that assumptions (A1) and (A2) hold, $w, \delta \in L^\infty(\mathbb{R}_+, \mathbb{R}^m)$ and $y_r, \nu \in W^{1,\infty}(\mathbb{R}_+, \mathbb{R}^m)$. Setting $\lambda_\nu := \lambda + \|\nu\|_{L^\infty(\mathbb{R}_+, \mathbb{R}^m)}$, Eq. (3.19) trivially implies that

$$\limsup_{t \rightarrow \infty} \|y_r(t) - y(t)\| \leq \lambda_\nu.$$

(2) In applications it is often useful to modify the gain adaptation law in (1.2) to

$$\dot{k}(t) = \begin{cases} \gamma(\|e(t)\| - \lambda)^2 & \text{if } \|e(t)\| \geq \lambda, \\ 0 & \text{if } \|e(t)\| < \lambda, \end{cases}$$

where $\gamma > 0$ is an additional design parameter. This modification with suitably chosen γ can be used to improve the transient response of the closed-loop system.

4. Applications to retarded systems and integrodifferential systems

In this section we show that Theorem 3.1 can be applied to retarded systems and integro-differential systems. We solve the adaptive λ -tracking problem for these classes of systems under the assumptions that the plant

is minimum-phase and that the high-frequency gain matrix has its spectrum in the open right-half plane. Moreover, it turns out that in both cases the internal variables of the controlled plant remain bounded.

4.1. Retarded systems

In the following we extend any function $F \in BV([a, b], \mathbb{R}^{n \times n})$ to the whole real axis by setting $F(t) = F(a)$ for $t < a$ and $F(t) = F(b)$ for $t > b$. Any measurable function $f : \Omega \rightarrow \mathbb{R}^n$, $\Omega \subset \mathbb{R}$, will be extended to the whole real axis by defining $f(t) = 0$ for $t \notin \Omega$. For $F = (F_{ij}) \in BV([0, h], \mathbb{R}^{n \times n})$ and $f = (f_1, \dots, f_n)^T$, $f_i \in L^1_{\text{loc}}(\mathbb{R}, \mathbb{R})$, $1 \leq i \leq n$, we define

$$dF * f := \begin{pmatrix} \sum_{j=1}^n dF_{1j} * f_j \\ \vdots \\ \sum_{j=1}^n dF_{nj} * f_j \end{pmatrix},$$

where dF_{ij} denotes the Borel measure on \mathbb{R} induced by F_{ij} and $dF_{ij} * f_j$ denotes the convolution of the measure dF_{ij} and the function f_j (on the whole real line). If f is continuous on $[-h, \infty)$, then, of course, $dF * f$ can be expressed as a Riemann–Stieltjes integral

$$(dF * f)(t) = \int_0^h dF(\tau) f(t - \tau) \quad \text{for } t \geq 0.$$

Consider the retarded system

$$\dot{x} = dA * x + Bu, \quad x|_{[-h, 0]} = x_0 \in C([-h, 0], \mathbb{R}^n), \tag{4.1a}$$

$$y = Cx, \tag{4.1b}$$

where $A \in BV([0, h], \mathbb{R}^{n \times n})$, $B \in \mathbb{R}^{n \times m}$ and $C \in \mathbb{R}^{m \times n}$. We assume that

$$\sigma(CB) \subset \mathbb{C}_0 \tag{4.2}$$

and

$$\det \begin{pmatrix} sI - \hat{A}(s) - B \\ C \quad 0 \end{pmatrix} \neq 0 \quad \text{for all } s \in \mathbb{C}_0^{cl}, \tag{4.3}$$

where $\hat{A}(s) := \int_0^h \exp(-s\tau) dA(\tau)$ denotes the Laplace–Stieltjes transform of A . The transfer function matrix $G(s)$ of Eqs. (4.1a) and (4.1b) is given by $G(s) = C(sI - \hat{A}(s))^{-1}B$.

Condition (4.2) is a generalization of the finite-dimensional relative-degree-one condition, whilst Eq. (4.3) is the so-called minimum-phase condition. As in the finite-dimensional case, see [5], it can be shown that Eq. (4.3) holds if and only if $G(s)$ has no zeros in \mathbb{C}_0^{cl} and the system satisfies the generalized Hautus conditions in \mathbb{C}_0^{cl} .

We show that if Eqs. (4.2) and (4.3) are satisfied, then the retarded system (4.1) can be written in the form (1.1) with assumptions (1) and (2) being satisfied. The condition (4.2) means in particular that CB is invertible. Therefore, $m \leq n$, $\dim \ker C = n - m$, $\dim \text{im } B = m$ and

$$\ker C \cap \text{im } B = \{0\}.$$

Let $v_1, \dots, v_{n-m} \in \mathbb{R}^n$ be a basis for $\ker C$, then the matrix

$$Q := (B(CB)^{-1}, v_1, \dots, v_{n-m})$$

is invertible, and moreover,

$$Q^{-1}B = \begin{pmatrix} CB \\ 0 \end{pmatrix}, \quad CQ = (I_m, 0).$$

It is useful to partition the matrix $Q^{-1}A(\cdot)Q$ as follows:

$$Q^{-1}A(\cdot)Q = \begin{pmatrix} A_{11}(\cdot) & A_{12}(\cdot) \\ A_{21}(\cdot) & A_{22}(\cdot) \end{pmatrix},$$

where $A_{11}(\cdot)$, $A_{12}(\cdot)$, $A_{21}(\cdot)$ and $A_{22}(\cdot)$ are matrices with entries in $BV([0, h], \mathbb{R})$ of size $m \times m$, $m \times (n - m)$, $(n - m) \times m$ and $(n - m) \times (n - m)$, respectively. Noticing that $Q^{-1}x$ is of the form

$$Q^{-1}x = \begin{pmatrix} Cx \\ \xi \end{pmatrix},$$

Eq. (4.1a) can be rewritten as

$$\dot{y} = dA_{11} * y + dA_{12} * \xi + CBu, \quad y|_{[-h, 0]} = y_0, \quad (4.4a)$$

$$\dot{\xi} = dA_{21} * y + dA_{22} * \xi, \quad \xi|_{[-h, 0]} = \xi_0, \quad (4.4b)$$

where

$$\begin{pmatrix} y_0 \\ \xi_0 \end{pmatrix} = Q^{-1}x_0.$$

Consider Eq. (4.4b) as differential equation in ξ with forcing term $dA_{21} * y$ and suppose for a moment that y can be chosen arbitrarily. For given initial functions ξ_0 and y_0 and given $y \in L^1_{loc}([-h, \infty), \mathbb{R}^m)$, let $\xi(t; \xi_0, y_0, y)$ denote the solution of the initial-value problem

$$\dot{\xi} = dA_{22} * \xi + dA_{21} * y; \quad \xi|_{[-h, 0]} = \xi_0, \quad y|_{[-h, 0]} = y_0.$$

Setting

$$\mathcal{H}(y) = dA_{12} * \xi(\cdot; 0, 0, y) + dA_{11} * y,$$

$$w = dA_{12} * \xi(\cdot; \xi_0, y_0, 0) + dA_{11} * y_0,$$

Eq. (4.4a) can be expressed as

$$\dot{y} = \mathcal{H}(y) + CBu + w, \quad y(0) = Cx_0(0).$$

Defining $\hat{A}_{22}(s) := \int_0^h e^{-s\tau} dA_{22}(\tau)$, the minimum-phase assumption (4.3) implies that

$$\det(sI - \hat{A}_{22}(s)) \neq 0 \quad \text{for all } s \in \mathbb{C}_0^{cl},$$

see [8]. This means that the zero solution of the retarded equation $\dot{\xi} = dA_{22} * \xi$ is exponentially stable, and hence the linear operator $y(\cdot) \mapsto \xi(\cdot; 0, 0, y)$ maps the space $L^p(\mathbb{R}_+, \mathbb{R}^m)$ boundedly into itself for all $p \in [1, \infty)$, and moreover, $\xi(\cdot; \xi_0, y_0, 0) \in L^\infty(\mathbb{R}_+, \mathbb{R}^m)$. Consequently, \mathcal{H} satisfies assumption (A2) and $w \in L^\infty(\mathbb{R}_+, \mathbb{R}^m)$. Also notice that, by Eq. (4.4b), if $y(\cdot)$ is bounded, then $\xi(\cdot)$, and hence $x(\cdot)$, are bounded.

Combining Theorem 3.1 with the above findings yields the following corollary.

Corollary 4.1. *Assume that Eqs. (4.2) and (4.3) are satisfied and let $\lambda > 0$ be given. Then, for all initial conditions $x_0 \in C([-h, 0], \mathbb{R}^n)$, $k_0 \in \mathbb{R}$, all $w \in L^\infty(\mathbb{R}_+, \mathbb{R}^m)$ and all reference signals $y_r \in W^{1, \infty}(\mathbb{R}_+, \mathbb{R}^m)$, the closed-loop system given by Eqs. (4.1a) and (4.1b) and (1.2) has the following properties:*

- (1) *The solution $(x(\cdot), k(\cdot))$ exists on $[0, \infty)$ and is unique,*
- (2) *$x(\cdot)$ is bounded and $\lim_{t \rightarrow \infty} k(t)$ exists and is finite,*
- (3) *$\limsup_{t \rightarrow \infty} \|y_r(t) - y(t)\| \leq \lambda$.*

4.2. Integrodifferential systems

Another interesting class of systems covered by Theorem 3.1 is a class of integrodifferential systems. Consider the system

$$\dot{x} = A * x + Bu, \quad x(0) = x_0 \in \mathbb{R}^n, \quad (4.5a)$$

$$y = Cx, \quad (4.5b)$$

where $A \in M(\mathbb{R}_+, \mathbb{R}^{n \times n})$, $B \in \mathbb{R}^{n \times m}$ and $C \in \mathbb{R}^{m \times n}$. The Volterra integrodifferential system

$$\dot{x}(t) = A_0 x(t) + \int_0^t A_1(t - \tau)x(\tau) d\tau + Bu(t), \quad x(0) = x_0 \in \mathbb{R}^n,$$

$$y(t) = Cx(t)$$

where $A_0 \in \mathbb{R}^{n \times n}$ and $A_1 \in L^1(\mathbb{R}_+, \mathbb{R}^{n \times n})$, is obviously a special case of Eqs. (4.5a) and (4.5b). As in Section 4.1 we assume that

$$\sigma(CB) \subset \mathbb{C}_0 \quad (4.6)$$

and

$$\det \begin{pmatrix} sI - \hat{A}(s) & -B \\ C & 0 \end{pmatrix} \neq 0 \quad \text{for all } s \in \mathbb{C}_0^{cl}, \quad (4.7)$$

where $\hat{A}(s) := \int_0^\infty \exp(-s\tau) dA(\tau)$.

Combining standard results from the theory of integrodifferential equations (see [3]) with ideas similar to those in Section 4.1, the following analogue of Corollary 4.1 for integrodifferential systems can be proved.

Corollary 4.2. *Assume that Eqs. (4.6) and (4.7) are satisfied and let $\lambda > 0$ be given. Then, for all initial conditions $x_0 \in \mathbb{R}^n$, $k_0 \in \mathbb{R}$, all $w \in L^\infty(\mathbb{R}_+, \mathbb{R}^m)$ and all reference signals $y_r \in W^{1,\infty}(\mathbb{R}_+, \mathbb{R}^m)$, the closed-loop system given by Eqs. (4.5a) and (4.5b) and (1.2) has the following properties:*

- (1) *The solution $(x(\cdot), k(\cdot))$ exists on $[0, \infty)$ and is unique,*
- (2) *$x(\cdot)$ is bounded and $\lim_{t \rightarrow \infty} k(t)$ exists and is finite,*
- (3) *$\limsup_{t \rightarrow \infty} \|y_r(t) - y(t)\| \leq \lambda$.*

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